

# Relaxed evolutionary power spectral density functions: A probabilistic approach to model uncertainties of non-stationary stochastic signals

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## Abstract

The identification of patterns and underlying characteristics of natural or engineering time-varying phenomena poses a challenging task, especially in the scope of simulation models and accompanying stochastic models. Because of their complex nature, time-varying processes such as wind speed, seismic ground motion, or vibrations of machinery in the presence of degradation oftentimes lack a closed-form description of their underlying evolutionary power spectral density (EPSD) function. To overcome this issue, a wide range of measurements exist for these types of processes. This opens up the path to a data-driven stochastic representation of EPSD functions. Rather than solely relying on time-frequency transform methods like the familiar short-time Fourier transform or wavelet transform for EPSD estimation, a probabilistic representation of the EPSD can provide valuable insights into the epistemic uncertainty associated with these processes. To address this problem, the evolutionary EPSD function is relaxed based on multiple similar data to account for these uncertainties and to provide a realistic representation of the time data in the time-frequency domain. This results in the so-called relaxed EPSD (REPSD) function, which serves as a modular probabilistic representation of the time-frequency content of stochastic signals. For this purpose, truncated normal distributions and kernel density estimates are used to

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2  
3 determine a probability density function for each time-frequency component. The REPSD  
4 function enables the sampling of individual EPSD functions, facilitating their direct appli-  
5 cation to the simulation model through stochastic simulation techniques like Monte Carlo  
6 simulation or other advanced methods. Even though the accuracy is highly dependant on  
7 the data available and the time-frequency transformation method used, the REPSD rep-  
8 resentation offers a stochastic representation of characteristics used to describe stochastic  
9 signals and can reduce epistemic uncertainty during the modelling of such time-varying pro-  
10 cesses. The method is illustrated by numerical examples involving the analysis of dynamic  
11 behaviour under random loads. The results show that the method can be successfully em-  
12 ployed to account for uncertainties in the estimation of the EPSD function and represent  
13 the accuracy of the time-frequency transformation used.

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24 *Keywords:* Evolutionary power spectral density function, Stochastic processes, Stochastic  
25 dynamics, Uncertainty quantification, Stochastic signals, time-frequency transformation  
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## 28 29 30 **1. Introduction**

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32 The description of natural phenomena in the context of simulation models is a chal-  
33 lenging problem. These phenomena such as wind and wave movements, seismic activities  
34 or climate changes are related to complex, interacting high-dimensional physical models.  
35 Also, engineering problems, such as vibrations of a component under changing material  
36 behaviour, can often only be modelled with an acceptable level of accuracy using complex  
37 models and experiments. Since these phenomena present time-varying properties in engi-  
38 neering they are often referred to as environmental processes or from a mathematical and  
39 modelling perspective, stochastic processes. In civil engineering, stochastic dynamics, and  
40 structural analysis, stochastic processes play a crucial role [1, 2, 3]. Stochastic dynamics is  
41 concerned with the study of probabilistic systems that evolve over time and has applications  
42 in structural reliability analysis. Specifically, the treatment of random vibrations is impor-  
43 tant in this field [4, 5, 6, 7]. To ensure the reliability of a structure, analysts need to consider

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4 14 the potential impact of environmental processes such as wind, wave, or seismic loads at the  
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6 15 design stage or carry out analyses for existing structures [8]. This often requires complex  
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8 16 simulations to accurately predict how the structure will respond to these processes and to  
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10 17 ensure that it meets safety requirements.

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12 18 In addition to the physical models, it is possible to introduce a complementary stochastic  
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14 19 model, which includes the formulation of suitable stochastic processes. To approximate  
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16 20 stochastic processes mainly three branches of approaches have been established up to now  
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18 21 [9]:

- 19 22 • The Karhunen-Lòeve (K-L) expansion, in which for engineering processes orthogonal  
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21 23 functions in time and space are combined linearly [10]. To describe the stochastic  
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23 24 processes, a formulation of the corresponding covariance functions must be available.  
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25 25 K-L can be used to simulate non-Gaussian and non-stationary stochastic processes  
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27 26 [11].
- 28 27 • Sampling representations, which are in the classical representation mostly suitable for  
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30 28 reproducing a full signal by deterministic samples available, however, methods like the  
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32 29 Withaker-Shannon interpolation require a limited bandwidth of the analysed processes  
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34 30 [12].
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36 31 • Spectral representation methods, based on the formulation of a power spectral density  
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38 32 (PSD) function [13].

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40 33 From an engineering perspective, when regarding environmental loads, formulating PSD  
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42 34 functions has advantages in vibration analysis, as they provide a method of directly char-  
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44 35 acterising the frequency content of stochastic signals. Artificial stochastic signals can be  
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46 36 considered as deterministic realisations of stochastic processes, whereas signals themselves  
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48 37 can also be measurements. Via PSD function estimation procedures, a signal can be decom-  
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50 38 posed into its harmonic components. In particular, the amplitudes and their distribution  
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52 39 over the frequencies are determined. However, to accurately calculate the PSD function  
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54 40 of a signal, certain mathematical conditions are necessary, such as dealing with continuous

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4 41 signals and signals of infinite length. Since these requirements cannot be met in practice,  
5 42 estimators are used.

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7 43 Over time, different spectral density estimators have been developed that offer certain  
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9 44 advantages and disadvantages. In the stationary case, common estimators such as the  
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11 45 periodogram [7], Welch's method [14], or Bartlett's method [15, 16] are widely employed.  
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13 46 For an overview, refer to [17]. These estimators all rely on the discrete Fourier transform  
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15 47 (DFT) [18]. Since signals of environmental processes often have a non-stationary character  
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17 48 which results in a frequency power change in time, so-called evolutionary PSD (EPSD)  
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19 49 functions need to be considered. These take into account the time-varying behaviour of  
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21 50 a signal and thus provide a more realistic representation in the resulting time-frequency  
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23 51 domain [19, 20, 21]. Different time-frequency transformation methods exist, with certain  
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25 52 advantages and disadvantages, particularly in the quality of the transformation as well as  
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27 53 the resolution in the time-frequency domain. An EPSD function can be estimated using,  
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29 54 among others, the short-time Fourier transform (STFT) [7, 22], wavelet methods [23, 24],  
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31 55 or the recently developed multi-taper S-transform (MTST) [25].

32 56 Once a suitable estimation of a PSD or EPSD is available, artificial stochastic signals  
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34 57 can be generated using the Spectral Representation Method (SRM), either in the stationary  
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36 58 case [13] or in the non-stationary case [26]. The latter is of interest in this work. For the  
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38 59 generation of stochastic signals, all stochastic simulation methods are applicable, such as the  
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40 60 widely used Monte Carlo (MC) simulation [27, 28] but also the usage of advanced techniques  
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42 61 such as subset simulation [29], line sampling [30], or directional importance sampling [31]  
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44 62 are possible.

45 63 However, determining these signals via a transformation method of choice and even  
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47 64 further utilise these representations for e.g. structural reliability analysis is challenging  
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49 65 due to the presence of uncertainties [32]. Uncertainties can be divided into aleatory and  
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51 66 epistemic uncertainties [33], while on the one side, aleatory uncertainties describe irreducible  
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53 67 stochastic conditions, epistemic uncertainties are referred to as reducible uncertainties. If  
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55 68 both types of uncertainties occur simultaneously and are not separable, they are called  
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57 69 hybrid uncertainties [34, 35]. The assessment of uncertainties in simulations, analyses, and

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4 70 engineering systems is ubiquitous. If uncertainties are incorrectly quantified, consequences  
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6 71 can be disastrous. For example, a building under given loads could suffer devastating damage  
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8 72 that severely compromises the structural reliability. Different approaches to quantify and  
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10 73 propagate uncertainties are available, such as precise probabilistic methods [36, 37], non-  
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12 74 probabilistic methods [38] or imprecise probabilities [39].

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15 75 In general, stochastic signals and specifically real data records are subject to uncertain-  
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17 76 ties. These can result, for example, from poorly calibrated sensors, measurement errors,  
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19 77 an insufficient amount of data, while damaged or failed sensors can result in records with  
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21 78 missing data. Furthermore, the presence of uncertainties is a direct result of the inherent  
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23 79 complexity found in natural environmental processes. In addition, due to the mentioned  
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25 80 mathematical requirements on the signal for the time-frequency transformations, only esti-  
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27 81 mates of the EPSD function can be determined, which leaves the epistemic uncertainty of  
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29 82 the EPSD functions unidentified. To overcome these issues, a range of PSD and EPSD func-  
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31 83 tion estimation techniques involving uncertainty quantification have been proposed. Missing  
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33 84 data problems are treated in [40, 41, 42], an interval-valued PSD function from similar data  
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35 85 has been derived in [43], while in [44] a set of accelerograms is analysed to derive reliability  
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37 86 bounds. An interval-valued signal can be transformed to an interval-valued PSD function  
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39 87 using the interval DFT transform [45]. In particular, the steadily growing databases, for  
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41 88 instance [46, 47, 48], contribute to a better understanding of environmental processes and  
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43 89 the quantification of uncertainties.

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46 90 In addition to the above mentioned approaches, the authors of this work derived a prob-  
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48 91 abilistic model of a set of similar PSD functions, the relaxed PSD function [49]. However,  
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50 92 the proposed methodology is only valid for stationary and Gaussian stochastic signals. The  
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52 93 non-stationary case needs to involve the EPSD function estimation for a more accurate rep-  
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54 94 resentation of environmental processes. Additionally, the spectral representation by SRM  
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56 95 and the stochastic simulation need to be reconsidered. In this work a modular framework for  
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58 96 the representation of non-stationary stochastic signals via EPSD functions using artificial  
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60 97 stochastic signals generated by the SRM is proposed. The EPSD function estimation will  
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62 98 be carried out using the recently developed MTST [25]. Once the ensemble is derived, each

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4 99 spectral density per time-frequency component will be transformed into a probability density  
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6 100 function. To underline the modularity of the construction of a REPSD function, simple trun-  
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8 101 cated normal distributions will be used as basis, additionally kernel density estimations are  
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10 102 implemented to construct custom probability density functions (PDF) in the time-frequency  
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12 103 space and to show, that arbitrary PDF types could be implemented. Both approaches have  
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14 104 their advantages and disadvantages, depending on the amount and appearance of the data.  
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16 105 The resulting REPSD function is used to perform numerical simulations of the dynamic  
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18 106 behaviour of systems subjected to environmental processes by sampling individual EPSP  
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20 107 functions applied to the model via MC simulation.

21 108 This work is organised as follows: Basic concepts of the SRM and EPSP function es-  
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23 109 tablishment, important for the remainder of this work will be explained in Section 2. In  
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25 110 Section 3 the methodology of constructing the REPSD function will be elaborated. The  
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27 111 obtained REPSD model will be validated by MTST estimations of EPSP functions and com-  
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29 112 pared to the source EPSP of an artificial environmental process. To illustrate the strengths  
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31 113 and advantages of the REPSD function, two different numerical examples are presented in  
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33 114 Section 4. The final conclusions and a critical discussion of the obtained results are given in  
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35 115 Section 5.

## 36 37 116 **2. Preliminaries**

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40 117 In this section necessary methodologies for the REPSD representation are introduced.  
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42 118 These mainly include the representation of non-stationary stochastic processes by a spectral  
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44 119 representation (Section 2.1), the estimation of EPSP functions from generated stochastic  
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46 120 signals by the state of the art MTST method (Section 2.2) and revisiting the kernel density  
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48 121 estimation (KDE), which is used to determine PDF representations during the REPSD  
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50 122 construction (Section 2.3).

### 51 52 53 123 *2.1. Representation of non-stationary stochastic processes*

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55 124 A convenient way to generate sample functions  $X(t)$  that represent non-stationary stochas-  
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57 125 tic processes in a time-domain is presented in [26]. For a source EPSP  $S_X(\omega, t)$  the spectral

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4 126 representation of non-stationary stochastic processes can be stated as

$$5  
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7 X(t) = \sqrt{2} \sum_{n=0}^{N_\omega-1} \sqrt{2S_X(\omega_n, t)\Delta\omega} \cos(\omega_n t + \varphi_n), \quad (1)$$

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10 127 in which,

$$11  
12 \omega_n = n\Delta\omega, \quad n = 0, 1, 2, \dots, N_\omega - 1, \quad (2)$$

$$13  
14 \text{and } \Delta\omega = \frac{\omega_u}{N_\omega},$$

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17 128  $t \in [0, T]$  is the time domain with  $T$  being the ultimate time.  $\omega_n$  are discretised frequency  
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19 values and  $\omega_u = f_s/4\pi$  is the upper cutoff-frequency determined by the time discretisation  
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21  $\Delta t$  and sampling ratio  $f_s = 1/\Delta t$ .  $\varphi_n$ , with  $n = 0, 1, \dots, N_\omega - 1$  are uniformly distributed  
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23 phase angles in the range  $[0, 2\pi]$ .  $N_t$  is the desired number of time instances, such that  
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25  $\Delta t = T/N_t$ .  $S_X(\omega_n, t)$  in Eq. (1) is the source EPSD and could be of arbitrary shape. A  
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27 133 non-separable EPSD as in [50] has been chosen as source EPSD:

$$28  
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30 S_X(\omega, t) = \left(\frac{\omega}{5\pi}\right)^2 \cdot \exp[-0.15t] \cdot t^2 \cdot \exp\left[-\left(\frac{\omega}{5\pi}\right)^2 t\right]. \quad (3)$$

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33 134 For a discretised time-frequency space this relation is depicted in Fig. 1. As already estab-  
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35 135 lished in the SRM [13], if  $N_\omega \rightarrow \infty$  it can be assumed that the sample function in Eq. (1)  
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37 136 presents an accurate simulation for a non-stationary stochastic process. Since only a limited  
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39 137 number  $N_\omega$  of approximation terms in Eq. (3) is feasible, the sample functions in  $X(t)$  are  
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41 138 referred to as stochastic signals or simply signals.

## 42 43 44 139 *2.2. Evolutionary power spectral density estimation*

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46 140 The generated sample functions in Eq. (1) are regarded as arbitrary stochastic signals  
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48 141 but with the same source EPSD. A challenging task remains to find a robust estimator for  
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50 142 this source EPSD when considering only a finite number of generated stochastic signals.  
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52 143 In [25] the MTST method, which exhibits a significant variance reduction in comparison to  
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54 144 other EPSD estimation procedures such as the Priestley method and wavelet-based methods,  
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56 145 has been presented. Given a non-stationary stochastic signal  $X(t)$  and  $M$  time-frequency

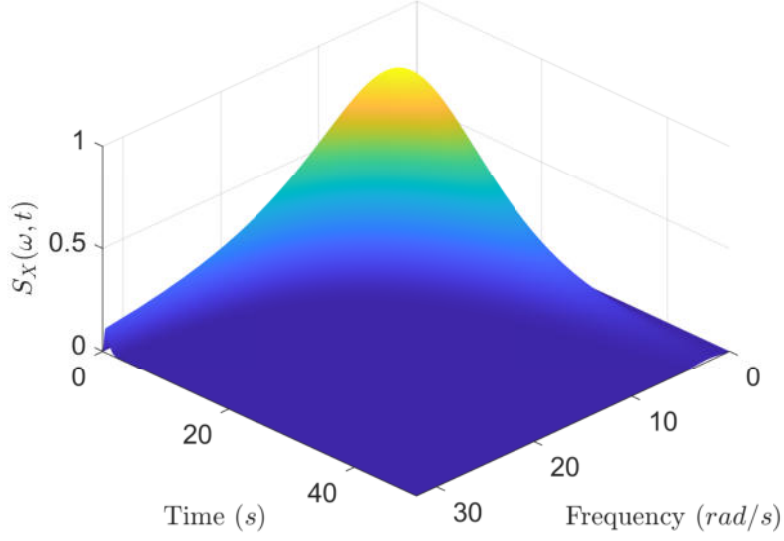


Figure 1: Non-separable EPSD function (Eq. (3)) for  $\omega \in [0, 32.17]$  rad/s and  $t \in [0, 50]$  s.

146 Hermite windows, denoted as  $\Psi_m(\omega, t)$ ,  $m = 0, 1, \dots, M - 1$ , with zero-padding  $X_0(t) =$   
 147  $\{0_1, \dots, 0_{N_t/2}, X(t), 0_1, \dots, 0_{N_t/2-1}\}$ , the so-called S-transform is given to be

$$s_m(\omega, t) = \sum_{k=-\underline{c}}^{\bar{c}} \Psi_m(\omega, k\Delta t - t) \cdot X_0(k\Delta t) \cdot \exp[-i2\pi\omega k\Delta t] \cdot \Delta t \quad (4)$$

148 in the case that  $\underline{c}, \bar{c}, \rightarrow \infty$ , Eq. (4) has the maximum accuracy dependant only on  $M$ . In this  
 149 study  $\underline{c} = \lceil N_t/2 \rceil + 1$  and  $\bar{c} = \lfloor N_t/2 \rfloor$ . Please note that with the zero-padding a continuous  
 150 window over the time-domain can be regarded, i.e.  $X_0(t)$  is treated as a periodic signal.

151 The Hermite windows are constructed as following

$$\begin{aligned} \Psi_0(\omega, t) &= \pi^{-1/4} \cdot \sqrt{w(\omega)} \cdot \exp[-1/2 \cdot w(\omega)^2 \cdot t^2], \\ \Psi_1(\omega, t) &= \sqrt{2}\pi^{-1/4} \cdot w(\omega)^{3/2} \cdot t \cdot \exp[-1/2 \cdot w(\omega)^2 \cdot t^2], \\ \Psi_m(\omega, t) &= \sqrt{2/m} \cdot w(\omega) \cdot t \cdot \Psi_{m-1}(t, \omega) - \sqrt{(m-1)/m} \cdot \Psi_{m-2}(t, \omega). \end{aligned} \quad (5)$$

152 With the shape function

$$w(\omega) = a \left[ 1 + \frac{b^2 \cdot |\omega/f_s|^{c+1}}{|b| \cdot |\omega/f_s|^{c+1} + 1} \right], \quad (6)$$

153 and parameters  $a, b, c$ . A detailed study on the choice of these parameters can be found in  
 154 [51]. With the full definition of the Hermite windows, the estimator for  $S_X(\omega, t)$  is established



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4 155 as

$$\hat{S}_X(\omega, t) = \frac{1}{M} \sum_{m=0}^{M-1} s_m^*(\omega, t) s_m^T(\omega, t). \quad (7)$$

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8 156 Here  $(\cdot)^*$  represents the conjugate operator, and  $(\cdot)^T$  the transpose operator.  
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11 157 *2.3. Kernel density estimation*  
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13 158 The KDE is a method of probability theory and statistics that allows to estimate the  
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15 159 PDF of a random variable without assuming a specific distribution. A brief explanation is  
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17 160 given here, while the reader is referred to [52, 53, 54] for a detailed explanation.

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19 161 A PDF is estimated by applying a kernel function to each individual data point. The  
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21 162 kernel function serves as a weighting factor to account for the contributions of each data  
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23 163 point to the estimated density. Summing up all kernel functions will result in the estimated  
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25 164 PDF. The choice of kernel function and its width is crucial for the accuracy of the estimate  
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27 165 because it determines the shape of the estimated PDF. Typical kernel functions include the  
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29 166 Gaussian distribution or the Epanechnikov function. The KDE is often used to gain an  
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31 167 understanding of the distribution of the data, or as a basis for further analysis.

32 168 The estimation of the PDF using KDE can be expressed by  
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$$\hat{f}(x) = \frac{1}{n} \sum_{i=1}^n \frac{1}{h} K_h \left( \frac{x - x_i}{h} \right), \quad (8)$$

39 169 where  $\hat{f}$  is the estimated probability density,  $K_h$  is the kernel function,  $h$  is the bandwidth  
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41 170 of the kernels,  $x$  is the point at which the PDF is estimated and  $x_i$  are the observations, i.e.  
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43 171 the available data points. In this work, a Gaussian kernel is utilised.

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45 172 The bandwidth has a significant impact on the quality of the resulting estimated PDF.  
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47 173 For example, setting the bandwidth too high can lead to an over-smoothed result, while  
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49 174 choosing a bandwidth that is too low can overweight individual data points, leading to an  
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51 175 under-smoothed PDF characterised by multiple sharp peaks. There are several ways to find  
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53 176 an optimal bandwidth for the kernels. Assuming that the PDF to be estimated is Gaussian  
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55 177 distributed, a well-known rule is Scott's rule [55], which incorporates the number of data  
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4 178 points  $n$  as well as their standard deviation  $\sigma$

$$5 \quad h_{opt} = \left( \frac{4\sigma^5}{3n} \right)^{1/5} \approx 1.06\sigma n^{-1/5}. \quad (9)$$

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9 179 Another rule is Silverman's rule [52], which takes into account the interquartile range  
10 180  $IQR$  in addition to the number of data points  $n$  and their standard deviation  $\sigma$

$$11 \quad h_{opt} = 0.9 \min \left( \sigma, \frac{IQR}{1.34} \right) n^{-1/5}, \quad (10)$$

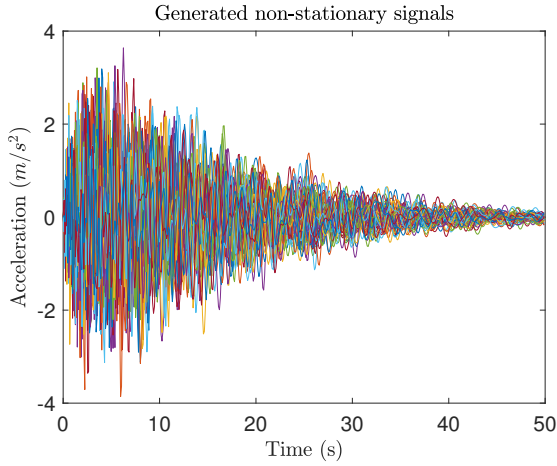
12  
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17 181 where  $IQR = Q_3 - Q_1$  with  $Q_1$  as lower quartile and  $Q_3$  as upper quartile.

### 18 19 20 21 182 **3. Methodology**

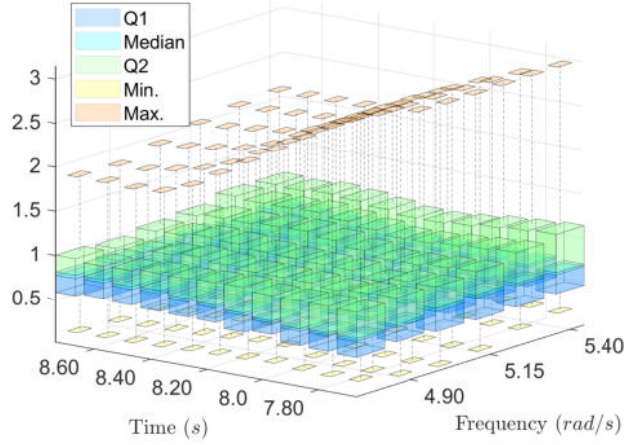
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23 183 In this section the novel estimation procedure for the REPSD function is presented.

#### 24 25 26 184 *3.1. Relaxed evolutionary power spectral density function*

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28 185 The novel construction of REPSD functions is particularly suitable when many records  
29 186 are available for particular phenomena. The aim is to reduce the epistemic uncertainty when  
30 187 describing stochastic signals. When records of environmental or natural processes such as  
31 188 wind loads, seismic ground motions, or other vibrations are available, it is still not deter-  
32 189 mined which EPSD function is a good approximation for this signal aggregation. For a  
33 190 thorough analysis, it is also interesting to examine whether these phenomena can be con-  
34 191 sidered directly related or not. The latter challenge is not part of this method and would  
35 192 require additional preliminary data analysis. However, from a modelling perspective, a com-  
36 193 mon characteristics formulation of environmental processes is essential to obtain simulation  
37 194 results. Furthermore, a stochastic, relaxed representation of EPSD functions for stochastic  
38 195 signals could lead to a new modelling perspective, as the representation of main features,  
39 196 such as first and second order moments, of natural processes can then be determined by  
40 197 statistical methods. Consider an ensemble that consists of a number of different EPSD esti-  
41 198 mations from stochastic signals. Fig. 2(a) shall represent signals from a natural process. The  
42 199 estimation procedure can be chosen according to the analyst's needs, throughout this work  
43 200 the presented MTST method in Section 2.2 with the estimator in Eq. (7) is used. This set of



(a) Non-homogeneous stochastic signals with  $N_\omega = 256$



(b) Ensemble  $\{\hat{S}_{X_i}\}, i = 1, 2, \dots, 50$  for  $t \in [7.6, 8.8]$  s and  $\omega \in [4.75, 5.4]$  rad/s

Figure 2: 50 source signals obtained from Eq. (1)(left), boxplot with full range and quartiles for the ensemble in  $\{\hat{S}_{X_i}\}$  (right).

EPSPD estimations with cardinality  $N_e$  can be stated as an ensemble:  $\{\hat{S}_{X_i}\}, i = 1, 2, \dots, N_e$ . Here  $N_e$  is determined by the number of regarded records available, containing either measurements or artificially generated stochastic signals (as in Fig. 2(a)). Each EPSPD function estimation, denoted by the index  $i$ , is discretised over the frequency and time domain, i.e.  $s_{i,\omega_n,t_k} = \hat{S}_{X_i}(\omega_n, t_k), n = 0, 1, \dots, N_\omega - 1, k = 1, \dots, N_t$ . The ensemble consists of the data-driven input collection of EPSPD estimations which lay the foundation of the stochastic input space for the REPSD function described below. In Fig. 2(b) the MTST estimations of the EPSPD functions are shown, the label abbreviations of the boxplot for each discretised frequency-time point represent the following: Min.: minimum value, Q1: lower quartile, Median: median value, Q2: upper quartile, Max.: maximum value of the data. It is obvious that the time-frequency transformation from stochastic signals delivers a massive amount of statistical data.

From the ensemble, following statistical moments are derived for each discretised point

214 of the EPSD ensemble

$$\mu_{\omega_n, t_k} = \frac{1}{N_e} \sum_{i=0}^{N_e-1} s_{i, \omega_n, t_k}, \quad (11a)$$

$$\sigma_{\omega_n, t_k} = \sqrt{\frac{1}{N_e} \sum_{i=0}^{N_e-1} (s_{i, \omega_n, t_k} - \mu_{\omega_n, t_k})^2}, \quad (11b)$$

215 with  $\omega_n = n\Delta\omega$ ,  $n = 0, 1, \dots, N_\omega - 1$ , and  $t_k = k\Delta t$ ,  $k = 1, 2, \dots, N_t$ . Eq. (11a) and  
 216 Eq. (11b) need to be calculated for each discretised frequency  $\omega_n$  and time instance  $t_k$ .  
 217 With this statistical information for each discretised point a probability density function  
 218 can be constructed. In theory, establishing arbitrary distribution types is possible. For the  
 219 sake of clarity, first, a simple distribution type is chosen. Assume a truncated normal (TN)  
 220 distribution for each point. These are then given to be

$$f_{\omega_n, t_k}^{TN}(s; \mu_{\omega_n, t_k}, \sigma_{\omega_n, t_k}, l_{\omega_n, t_k}, u_{\omega_n, t_k}) = \frac{1}{\sigma_{\omega_n, t_k}} \frac{\phi\left(\frac{s - \mu_{\omega_n, t_k}}{\sigma_{\omega_n, t_k}}\right)}{\Phi\left(\frac{u_{\omega_n, t_k} - \mu_{\omega_n, t_k}}{\sigma_{\omega_n, t_k}}\right) - \Phi\left(\frac{l_{\omega_n, t_k} - \mu_{\omega_n, t_k}}{\sigma_{\omega_n, t_k}}\right)}, \quad (12)$$

221 and  $\phi(\eta) = \frac{1}{\sqrt{2\pi}} \exp(-\frac{1}{2}\eta^2)$  is the standard normal distribution,  $\Phi(\zeta) = \frac{1}{2} (1 + \text{erf}(\zeta/\sqrt{2}))$  is  
 222 the corresponding cumulative distribution function. The lower and upper truncation bounds  
 223 are given to be  $l_{\omega_n, t_k}$  and  $u_{\omega_n, t_k}$ . The influence of the truncation bounds has been discussed for  
 224 the frequency domain in [49]. In this study the configuration  $l_{\omega_n, t_k} = 0$  and  $u_{\omega_n, t_k} = 2\mu_{\omega_n, t_k}$   
 225 proved to be more robust. The TN distribution yields a smooth representation of the EPSD  
 226 function values for each discretised point. Thus, outliers or gaps in the EPSD function value  
 227 are weighted less.

228 In some cases, however, it may be useful to represent the data more in its natural  
 229 appearance. In such a case, the KDE provides a much more inclusive representation of the  
 230 data. This can result in multiple peaks instead of smooth curve as in the case of the TN  
 231 distribution. If the data set is to be represented by KDE, Eq. (8) can be reformulated into  
 232 its KDE-driven REPSD form

$$\hat{f}_{\omega_n, t_k}^{KDE}(x; s_{i, \omega_n, t_k}, h_{\omega_n, t_k}) = \frac{1}{N_e} \sum_{i=1}^{N_e} \frac{1}{h_{\omega_n, t_k}} K_{h_{\omega_n, t_k}} \left( \frac{x - s_{i, \omega_n, t_k}}{h_{\omega_n, t_k}} \right). \quad (13)$$

1  
2  
3  
4 233 This equation is evaluated for each time-frequency point  $(\omega_n, t_k)$ . The parameter  $h_{\omega_n, t_k}$  de-  
5  
6 234 scribes an adaptive bandwidth for the KDE, which will be determined according to Eq. (10)  
7  
8 235 and the respective available EPSD values at the corresponding time-frequency point.

9 236 For both types of distribution functions it can be argued why they are superior, but this  
10  
11 237 always depends on the individual case, as the form and number of data are important for  
12  
13 238 this assessment. For a high number of data, a truncated normal distribution may be better,  
14  
15 239 as this may well lead to a smooth representation. Also, if a clear interval of the ensemble  
16  
17 240 data can be identified the TN approach would be better because this would make the choice  
18  
19 241 of truncation bounds for the PDF definition easier. If there is less data or multiple PDF  
20  
21 242 peaks are expected, a KDE may provide better results, since gaps and multiple peaks, for  
22  
23 243 instance, can be represented easier. However, here over-fitting could appear, in these cases  
24  
25 244 a careful examination of the sampled REPSDs about the ensemble and the underlying data  
26  
27 245 needs to be carried out. Special consideration to outliers must be given, in particular for  
28  
29 246 KDE approaches, because again here an over representation of outlier data could occur.  
30  
31 247 Note that the distributions presented are only two possibilities that could be used for the  
32  
33 248 REPSD model. Additionally, no correlations or dependencies have been considered so far.  
34  
35 249 To obtain a realisation of a truncated normal REPSD (TN-REPSD) named  $TS_X(\omega_n, t_k)$ , for  
36  
37 250 each  $\omega_n$  and  $t_k$  a sample is generated from the respective distributions in Eq. (12). To obtain  
38  
39 251 a realisation of a kernel density estimated REPSD (KDE-REPSD) named  $KS_X(\omega_n, t_k)$ , for  
40  
41 252 each  $\omega_n$  and  $t_k$  a sample is generated from the respective distributions in Eq. (13). In a  
42  
43 253 sense, Eq. (12) and Eq. (13) can be seen as the description of uncorrelated random fields.

### 44 254 *3.2. Optimised MTST parameters & error estimates*

45  
46  
47 255 The MTST estimation for each signal in Fig. 2(a) is dependant on the choice of the  
48  
49 256 parameters  $a, b, c$  of the window function in Eq. (6) and the number of Hermite window  
50  
51 257 orders  $M$  used. Since the source EPSD for the regarded signals is known as in Eq. (3), it  
52  
53 258 is possible to formulate an objective function. The objective function, also later on used as  
54  
55 259 error estimate, is formulated according to the Frobenius norm of matrices. Respectively for  
56  
57 260 two arbitrary EPSDs a residual matrix of the comparison of the two is stated as  $S_{res}(\omega_n, t_k) =$

1  
2  
3  
4 261  $S_1(\omega_n, t_k) - S_2(\omega_n, t_k)$  for  $n = 0, 1, \dots, N_\omega - 1, k = 1, \dots, N_t$ . For this residual matrix the  
5  
6 262 Frobenius norm is used to formulate an objective function and error estimate by

$$||S_{res}(\omega_n, t_k)||_F = \sqrt{\sum_{i=0}^{N_\omega-1} \sum_{j=1}^{N_t} |S_{res}(\omega_i, t_j)|^2}. \quad (14)$$

7  
8  
9  
10  
11  
12 263 Following constraints for the MTST parameters are introduced  $0.001 \leq a \leq 1$ ,  $0.001 \leq$   
13  
14 264  $b \leq 30$ ,  $0.001 \leq c \leq 1$ , and  $M \in 1, 2, \dots, 10$ , these constraints were chosen by the authors  
15  
16 265 after assessing the parameter discussion in [51]. For the minimisation of the residual matrix  
17  
18 266 a genetic algorithm optimisation was used as established in [56]. Following objective is  
19  
20 267 minimised once before the REPSD is constructed:

$$\arg \min_{a,b,c,M} \left\{ ||S_{res0}(\omega_n, t_k, a, b, c, M)||_F \right\}, \quad (15)$$

21  
22  
23  
24  
25  
26 268 where in this specific case  $S_{res0}(\omega_n, t_k) = S_X(\omega_n, t_k) - \overline{\hat{S}_X(\omega_n, t_k)}$ , which defines the Frobenius  
27  
28 269 norm between the source EPSD and the mean of one a priori chosen ensemble. The formal  
29  
30 270 calculation of the first and second order moments for the ensemble data can be found in  
31  
32 271 Appendix A.

### 33 34 35 272 *3.3. Error analysis*

36  
37 273 In Fig. 3, a single MTST estimation of one arbitrarily chosen signal in Fig. 2(a) is  
38  
39 274 depicted. In Fig. 4, a single realisation of the TN-REPSD is shown. Whilst the MTST  
40  
41 275 estimation is showing gaps for specific parts of the EPSD, the by TN-REPSD generated  
42  
43 276 sample EPSD is resembling the source EPSD in Fig. 1.

44  
45 277 The mean of the MTST estimations for all ensemble members is robust and yields a  
46  
47 278 good approximation. But a single TN-REPSD sample, as in Fig. 4, shows in comparison to  
48  
49 279 a single MTST estimation a better resemblance to the source EPSD due to the construc-  
50  
51 280 tion of a non-correlated random field that describes the ensemble's statistics in frequency  
52  
53 281 and time domain. For all error evaluations, the direct absolute differences between two  
54  
55 282 EPSD functions, differences between the ensemble's mean and REPSD sample's mean, or  
56  
57 283 the equivalent difference for the standard deviation are regarded. Also, the errors according

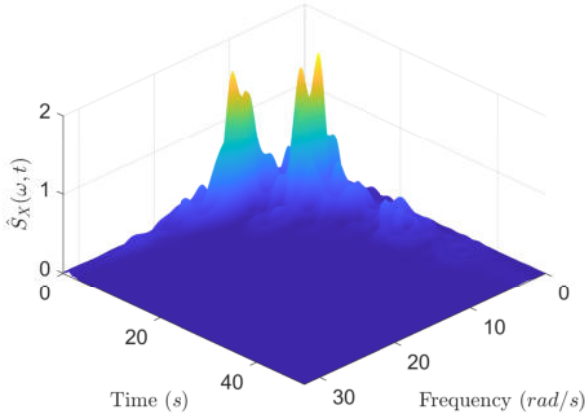


Figure 3: Single MTST estimation for one signal as depicted in Fig. 2(a).

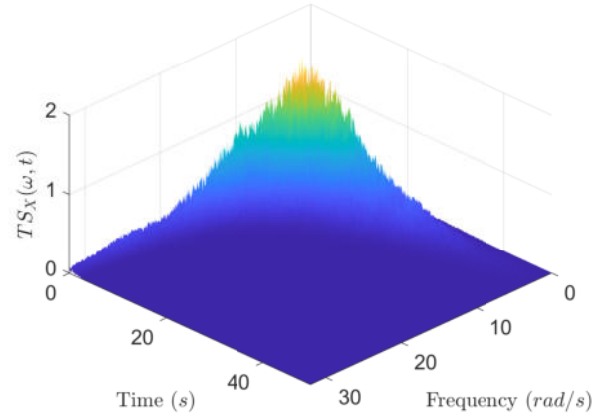


Figure 4: Single TN-REPSD sample generated according to Eq. (12).

to the Frobenius norm introduced in Eq. (14) are shown. The error and convergence analysis for the TN-REPSD representation and sampling has been performed. The results are shown in Fig. 5, here following notations were used: The source EPSD function is described as  $S_X(\omega, t)$ , the ensemble mean estimated by MTST is  $\overline{\hat{S}_X(\omega_n, t_k)}$ , the mean of generated TN-REPSD samples is denoted as  $\overline{TS_X(\omega, t)}$ .

In Fig. 5(a) the error between the source EPSD and the mean of the MTST estimates is shown. The maximum deviation amounts to  $\approx 0.2$ . This error is the reference, since it measures the deviation of the MTST estimation to the source EPSD function. In Fig. 5(b) the error between the mean of  $N_{10k} = 10000$  generated TN-REPSD samples and the source EPSD is shown. Here it can be seen that the error is similar to the error of the MTST estimation mean. This leads to the conclusion that the constructed TN-REPSD is mainly dependant on the quality of the EPSD estimation.

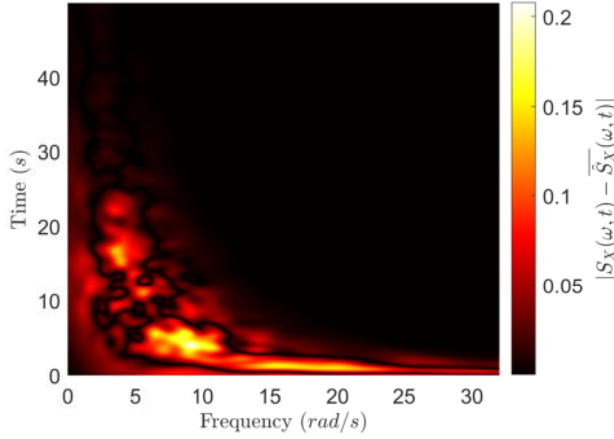
TN-REPSD is not adding any errors but offering a fully stochastic description of an estimated EPSD. In Fig. 5(c) the error between the mean of the MTST estimated ensemble is compared to the mean of the  $N_{10k}$  generated TN-REPSD samples. From the three results Fig. 5(a) - 5(c), the error map in Fig. 5(c) exhibits the smallest error. This is expected, because the TN-REPSD is constructed out of the  $N_e = 50$  ensemble members. In Fig. 5(d)

301 a convergence analysis based on the MC method has been carried out for different number  
 302 of generated TN-REPSD samples  $N \in \{1, 2, 2.5, 3, 4.5, 4, 4.5, 5\} \cdot 10^4$ . For larger numbers of  
 303  $N$ , the error between the TN-REPSD mean and the ensemble mean is decreasing (blue line,  
 304 blue axis), the error between the source EPSD and the mean of the TN-REPSD is constant  
 305 (orange line, orange axis). The dashed black line refers to the direct error of the source EPSD  
 306 function in comparison to the mean of the MTST ensemble in the Frobenius norm. This  
 307 error serves here as a reference and corresponds to the orange axis. The orange line is below  
 308 this value. All these results in Fig. 5 indicate that the TN-REPSD does not introduce any  
 309 additional error and delivers a good representation of the provided input ensemble (data).  
 310 Additionally, the convergence of the MC simulations follows the law of large numbers, i.e.  
 311 with a larger sample size the error reduces. Similar considerations are carried out for the  
 312 KDE-REPSD in Fig. 6.

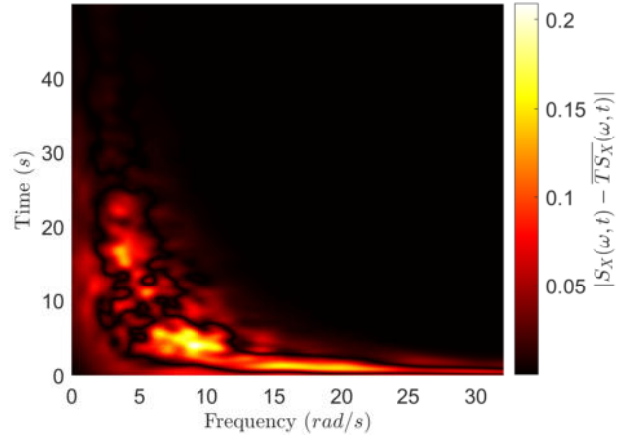
313 In Fig. 6, the notations are equivalent to the previous graph,  $\overline{KS_X(\omega, t)}$  denotes the  
 314 mean of the by KDE-REPSD generated EPSD samples. The general trend of these error  
 315 and convergence results are similar. Which again indicates that the stochastic representation  
 316 of the EPSD ensemble is mainly reliant on the EPSD estimation procedure. Please note  
 317 that a different ensemble has been used for the KDE-REPSD analysis, resulting in a slightly  
 318 different error in Fig. 6(a) compared to Fig. 5(a). In Fig. 6(c) higher local errors are observed.  
 319 Especially when comparing with the previous results of the TN-REPSD in Fig. 5(c). Also  
 320 for the KDE-REPSD generated samples it can be observed, that the dashed black line,  
 321 which is the reference error value for the source EPSD estimation, is always larger than  
 322 the orange line. This leads to the conclusion that no additional error is introduced by the  
 323 KDE-REPSD.

324 Additionally, for specific time instances  $t \in \{0.88 \text{ s}, 9.86 \text{ s}, 19.92 \text{ s}\}$  the EPSD function's  
 325 frequency space is analysed. The REPSD functions first and second order moment, that are  
 326 used for the construction of the REPSD (input for the distribution functions) are denoted by  
 327  $\mu[TS_X(\omega, t)]$  and  $\mu[KS_X(\omega, t)]$ , the mean parameter and  $\sigma(TS_X(\omega, t))$  and  $\sigma(KS_X(\omega, t))$ ,  
 328 the standard deviation parameter. For the MC sampling of the TN-REPSD and KDE-  
 329 REPSD the same  $N_{10k}$  samples from the previous results were used, the mean of the samples

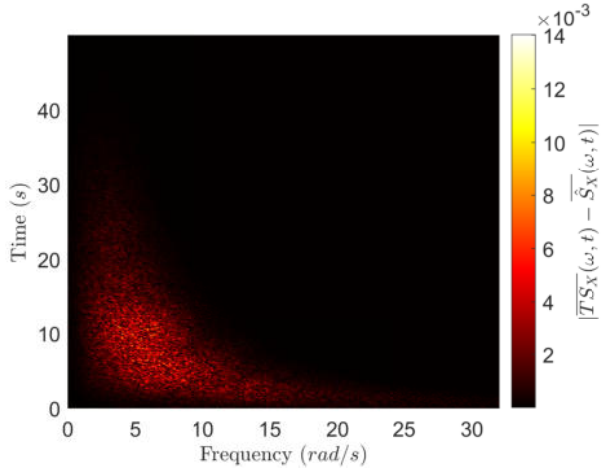




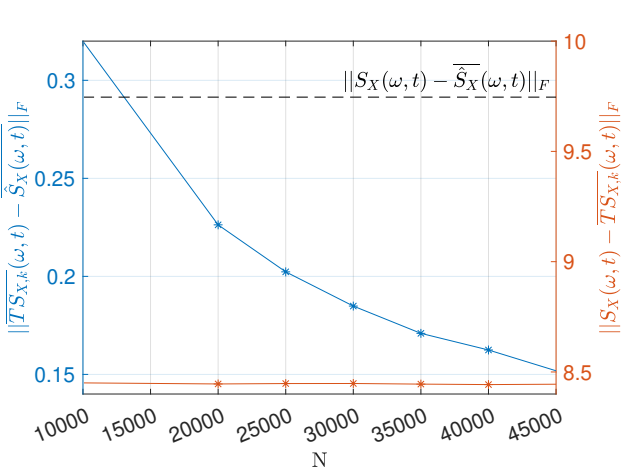
(a) Difference of source EPSP  $S_X(\omega, t)$ , vs. the mean of the MTST estimations  $\hat{S}_X(\omega_n, t_k)$ .



(b) Difference of source EPSP and the mean of TN-REPSD generated samples  $\overline{TS}_X(\omega, t)$ .

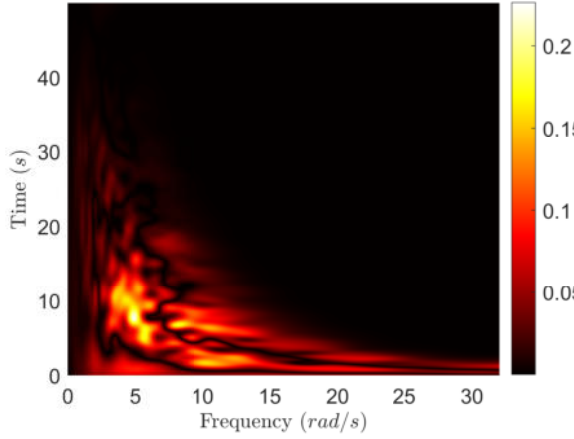


(c) Difference of the mean of the MTST estimation and the mean of TN-REPSD samples.

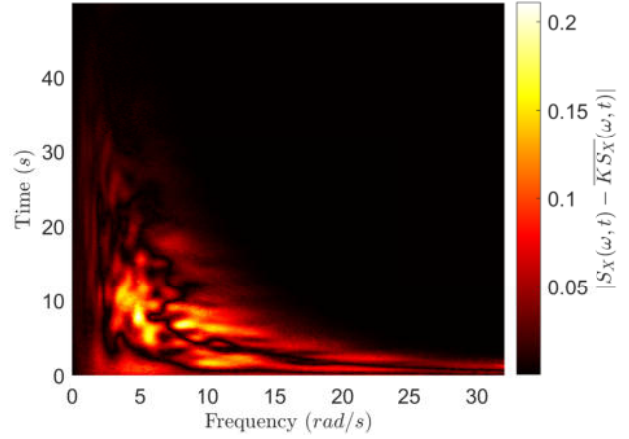


(d) Three Frobenius norm residuals, the black constant value belongs to the right (orange) axis.

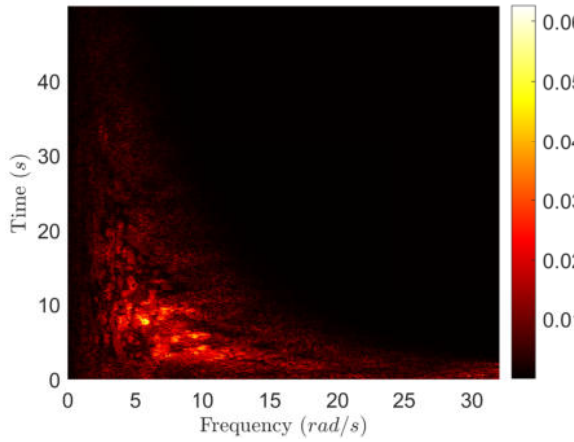
Figure 5: Errors, differences and convergence analysis for the TN-REPSD samples.



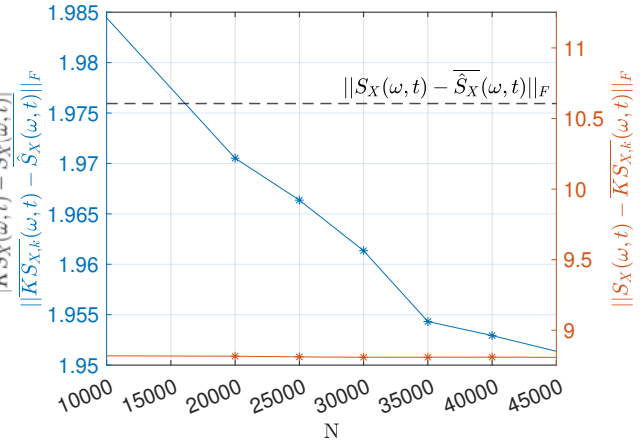
(a) Difference of source EPSP  $S_X(\omega, t)$ , vs. the mean of the MTST estimations  $\hat{S}_X(\omega_n, t_k)$ .



(b) Difference of source EPSP and the mean of KDE-REPSD generated samples  $\overline{KS_X}(\omega, t)$ .



(c) Difference of the mean of the MTST estimation and the mean of KDE-REPSD samples.



(d) Three Frobenius norm residuals, the black constant value belongs to the right (orange) axis.

Figure 6: Errors, differences and convergence analysis for the KDE-REPSD samples.

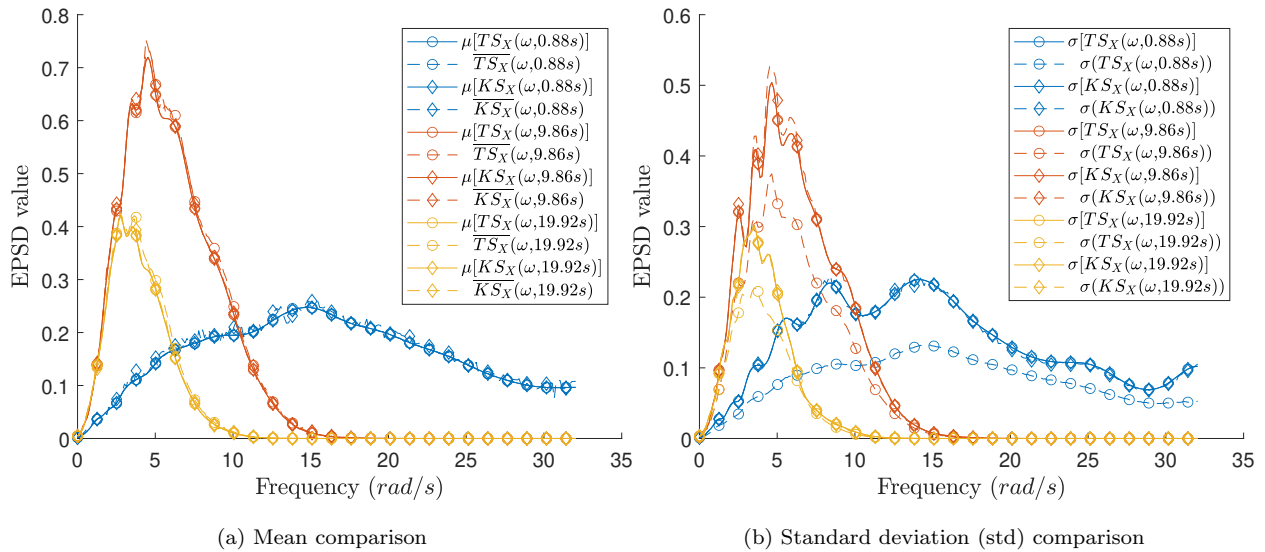


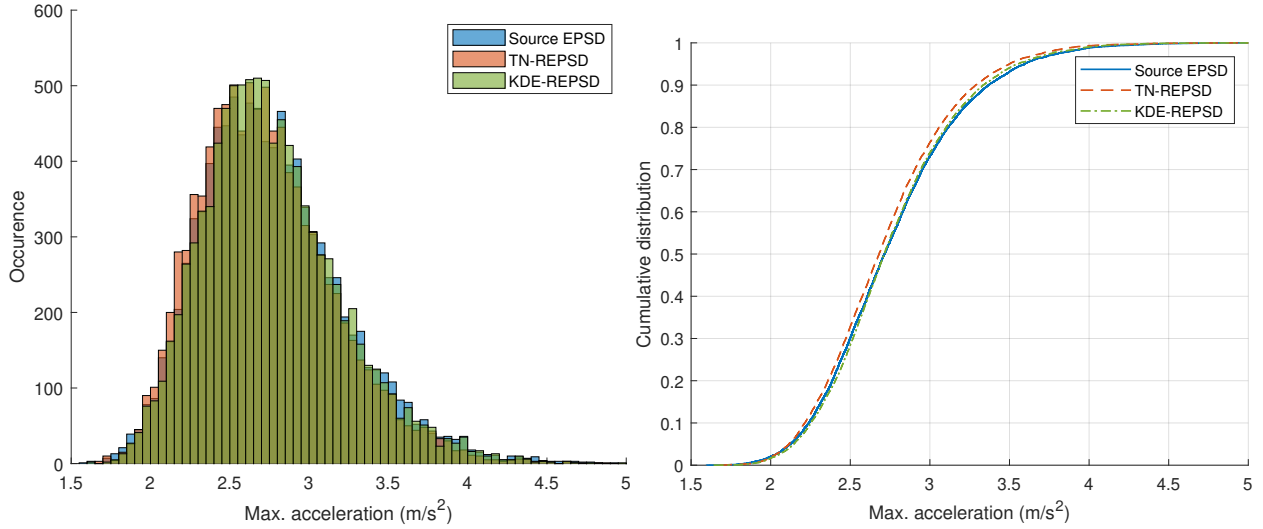
Figure 7: Stochastic modelling parameters in comparison to the MC samples statistics.

is denoted by  $\overline{TS_X(\omega, t)}$  and  $\overline{KS_X(\omega, t)}$ , the standard deviation of the samples is denoted by  $\sigma(TS_X(\omega, t))$  and  $\sigma(KS_X(\omega, t))$ . The mean over the frequency space Fig. 7(a) and the standard deviation over the frequency space Fig. 7(b) were compared.

### 3.4. Relaxed non-homogeneous spectral representation method

In this section the influence of the REPSD models to the generated time signals is evaluated. For this purpose, samples from both TN-REPSD (Eq. (12)) and the KDE-REPSD (Eq. (13)) are drawn and time signals are generated by using Eq. (1). A total of 10,000 EPSPD samples and corresponding signals are compared to each other. As a reference value, time signals generated from the source EPSPD given in Eq. (3) are used. The quantity considered for comparison is the absolute maximum value  $\max(|\ddot{x}(t)|)$  of the respective time signal  $\ddot{x}(t)$ , which represents an earthquake ground motion in this case.

The respective results are given in Fig. 8, where the histograms of the three respective cases are given in Fig. 8(a) and the corresponding empirical CDF are depicted in Fig. 8(b). It can be clearly seen that the time signals generated by the three models result in a very similar behaviour in terms of maximum acceleration. All histograms show a similar shape and distribution of the maximum values, which consequently is also visible in the empirical



(a) Histogram of the maximum acceleration

(b) Empirical CDF of the maximum acceleration

Figure 8: Histogram and empirical CDF of the maximum acceleration of the generated signals from the sampled EPSD.

CDFs. It should further be noted, that none of the models has some extreme values in any direction, thus the results can be considered of equal quality. However, minor differences can be seen, which may result from the influence of the random variables used in the stochastic process generation.

### 3.4.1. Energy of generated signals

As a further criterion for comparison, the total energy in the generated signals is considered. This analysis was carried out again for both the TN-REPSD (Eq. (12)) and the KDE-REPSD (Eq. (13)). The total energy of the signal can be determined by the following expression

$$E = \sum_{n=0}^{N_t} |x(n)|^2. \quad (16)$$

The analysis has been carried out with the identical signals generated in the previous section. In Fig. 9 an overview of the energy content of the signals is given. For better comparability, the energy content of the individual signals is arranged in ascending order. It can be clearly seen that all models result in signals with similar energy content and energy distribution.

In addition, Table 1 shows the min, max, mean and median for the 10,000 generated

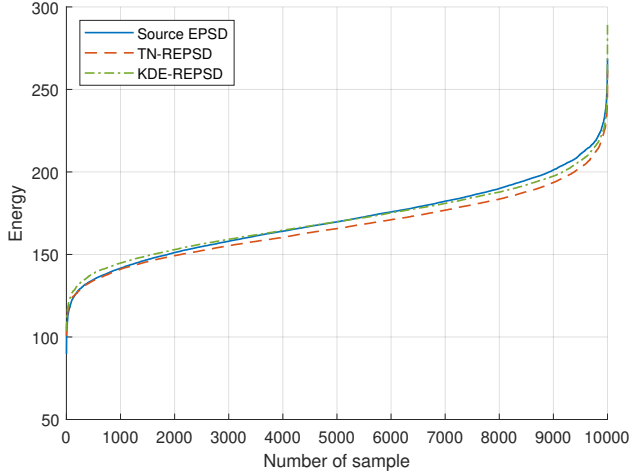


Figure 9: Energy of the generated signals in ascending order of energy content.

Table 1: Comparison of the energy content of the generated signals from the source EPSD, TN-REPSD and KDE-REPSD. Units are given in  $[m^2 s^{-4}]$ .

	Source EPSD	TN-REPSD	KDE-REPSD
min	89.7147	100.771	103.5557
max	268.484	261.6281	290.2797
mean	170.8195	166.7248	170.7418
median	169.7479	165.661	169.8419

signals. The min value of the source EPSD seems to be slightly smaller than for both relaxed models and the max value of the KDE-REPSD is also somewhat higher than the other values. However, since these are the extreme values, such a behaviour can be expected. In addition, those outlier values may only be reached by a very small portion of samples. In this comparison, the mean and median are more meaningful as often a high number of samples is applied to the system under investigation. For these two values, it is clear to see that they are in the similar range. Thus, it can be concluded that the signals have an identical energy content, at least in an averaged sense.

## 4. Numerical examples

To show possible applications of the REPSD approach, two numerical examples are investigated in this work.

### 4.1. Finite element model of a steel frame structure with irregular mass

Underlining the versatility of the REPSD approach, the seismic response of a steel moment resisting frame Finite Element Model (FEM) is analysed. The FEM is inspired by the results presented in [57], here a numerical model of a low rise moment steel frame with an irregular mass on the top storey has been presented and validated. In Fig. 10(a) the steel frame's dimensions, nodes and elements are depicted. White dots represent the conjunction nodes, the black dot represents a mass node. The beams and columns are consisting of H-shaped fibre steel material elements which are implemented displacement-based via the Open System for Earthquake Engineering Simulation (OpenSees) [58, 59]. Following length quantities are defined:  $h_0 = 680\text{mm}$ ,  $h_1 = 630\text{mm}$ ,  $W = 1600\text{mm}$ ,  $b = 800\text{mm}$ ,  $b_F = 45\text{mm}$ ,  $h_T = 100\text{mm}$ ,  $b_S = 6\text{mm}$ ,  $h_F = 8\text{mm}$ . The material properties of the steel is characterised by the skeleton stress-strain curve, which defines compression and tension points, see Fig. 10(b). Within OpenSees the uniaxial material with hysteretic properties has been chosen. The stress-strain points in Fig. 10(b) define the envelope. The model's columns are fixed within the foundation assuming a damping ratio of 0.02.

For the seismic ground motion, the SRM and both REPSD approaches with the relaxed SRM approach have been tested. The source EPSD as in Eq. (3) has been chosen, the SRM and relaxed SRM generated processes are scaled down by a factor of 10. MC simulation results of the SRM generated signals are depicted in Fig. 11(a), the generated signals serve as artificial seismic ground motion signals applied to the FEM. The displacements of the FEM's 4-th storey centre node has been chosen as quantity of interest. Exemplary results for the displacement are shown in Fig. 11(b). Each artificial seismic ground motion again leads to a different EPSD estimation, as respectively depicted in Fig. 12(a). In Fig. 12(b) respectively 3 different response EPSDs of the 50 generated ensemble members are shown.

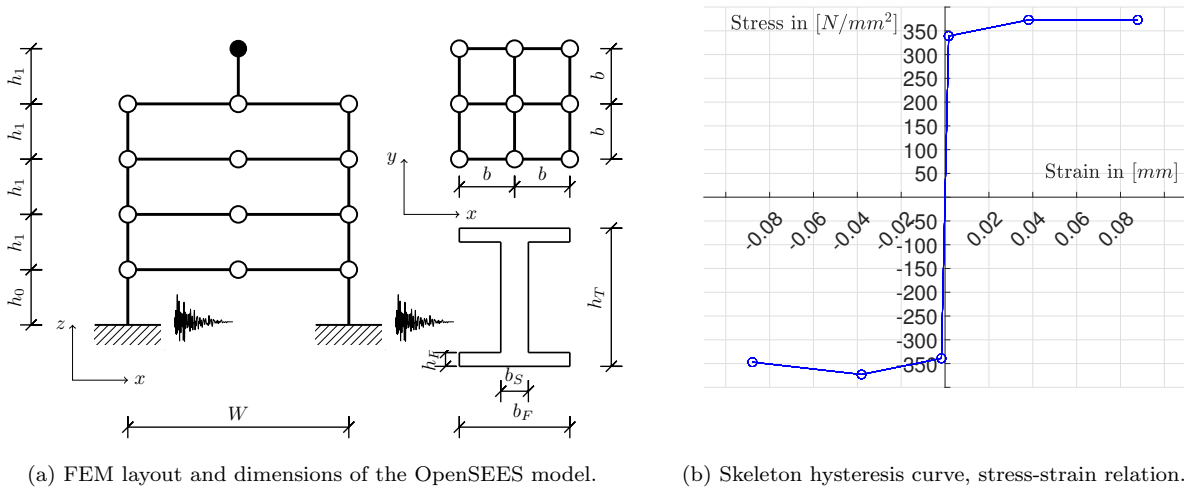


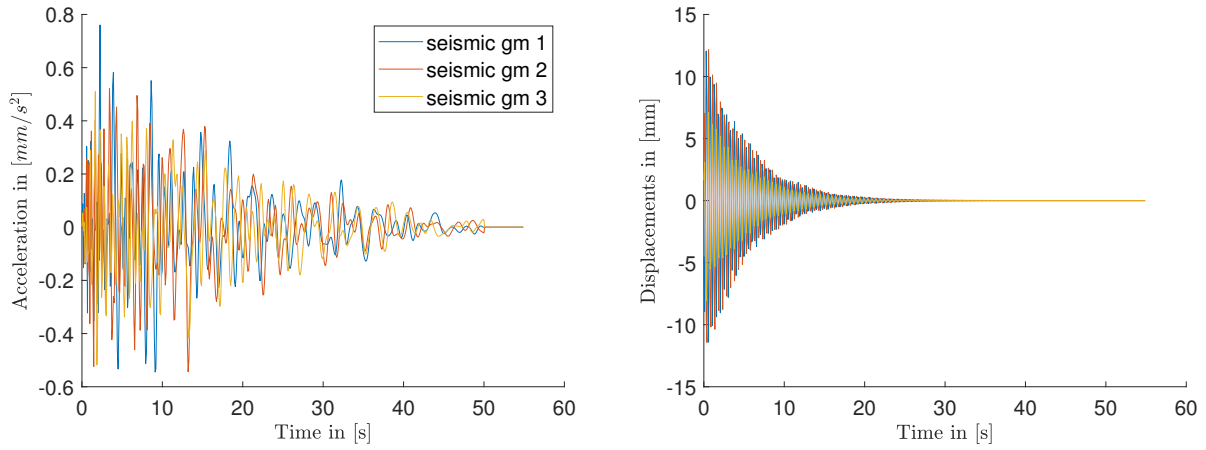
Figure 10: OpenSEES FEM dimensions and stress-strain relations of the material.

Thus the FEM input as well as output are treated as signals from which EPSDs ensembles are constructed.

For the REPSD generation, in each case the same input signal and therefore EPSD ensemble with 50 members has been used. The number of 50 reflects the limited availability of data, in this case artificial ground motion records. The same ensemble for each estimation has been used in order to ensure comparability between the TN-REPSD and KD-REPSD approach. Both REPSD models are then compared with results of 5000 MC simulations. The procedure for the input signal, is already validated in the preliminaries. Only a linear scaling was applied to the signals, which only leads to smaller EPSD function values but no change of the time-frequency components.

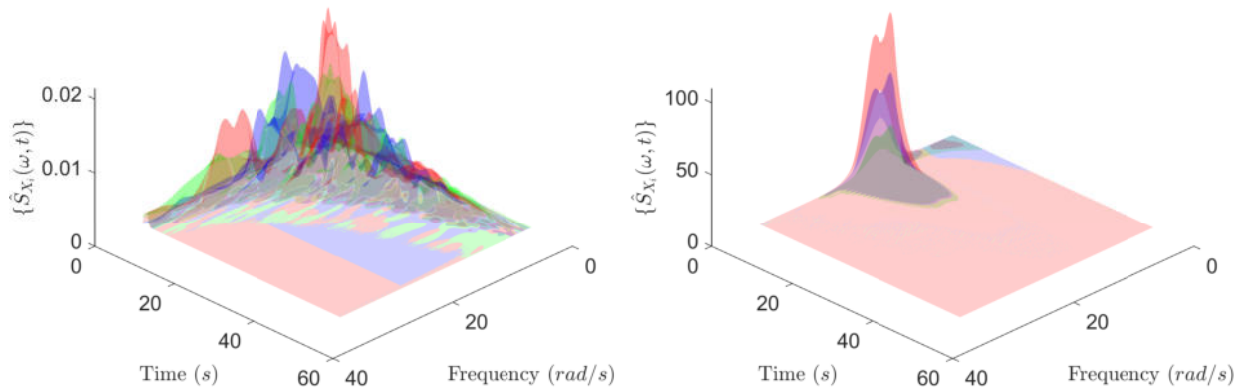
First a comparison of the full time history of the FEM response is carried out, the results are shown in Fig. 13(a). A closer comparison of the results for a time interval of 2s is shown in Fig. 13(b). In these results it can be seen that the stochastic model of the TN- and KDE-REPSD ground motion data is representing the stochastic properties of the process with satisfying accuracy.

During the analysis of the FEM, no closed-form solution of the response EPSD is available, therefore the comparison with the MC simulation is considered as benchmark. To



(a) Three artificial seismic ground motion signals (obtained by SRM). (b) Corresponding three response results of the FEM's top centre node.

Figure 11: FEM frame structure model input and response. The abbreviation *gm* refers to ground motion.



(a) Three deterministic EPSPD estimations by MTST for the signals in Fig. 11(a). (b) Corresponding three response EPSPD estimations by MTST for Fig. 11(b).

Figure 12: EPSPD estimations by MTST for model input (artificial seismic ground motion) and output (displacement responses).



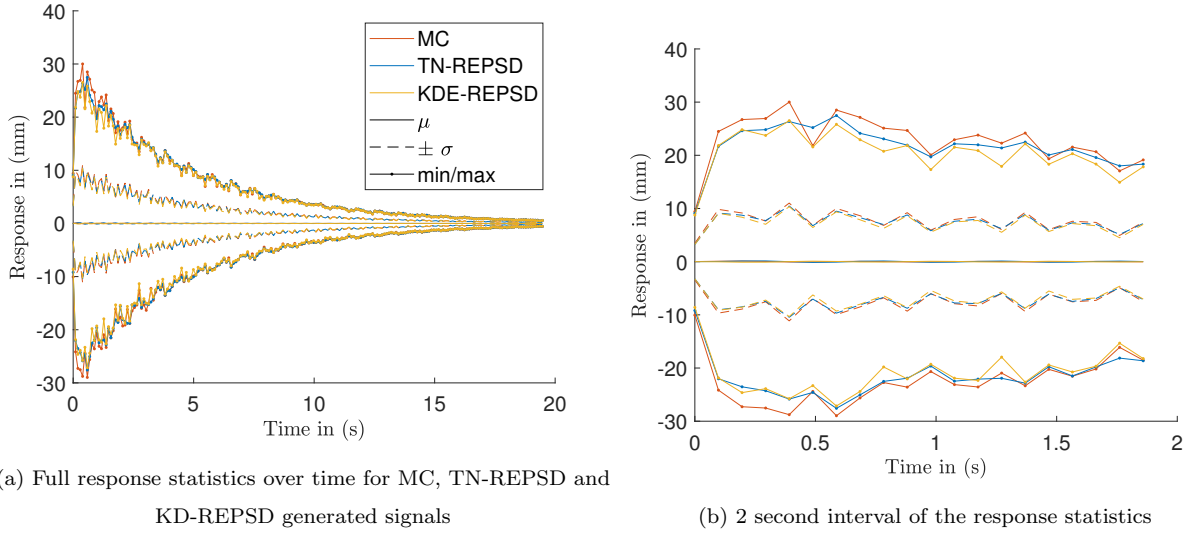
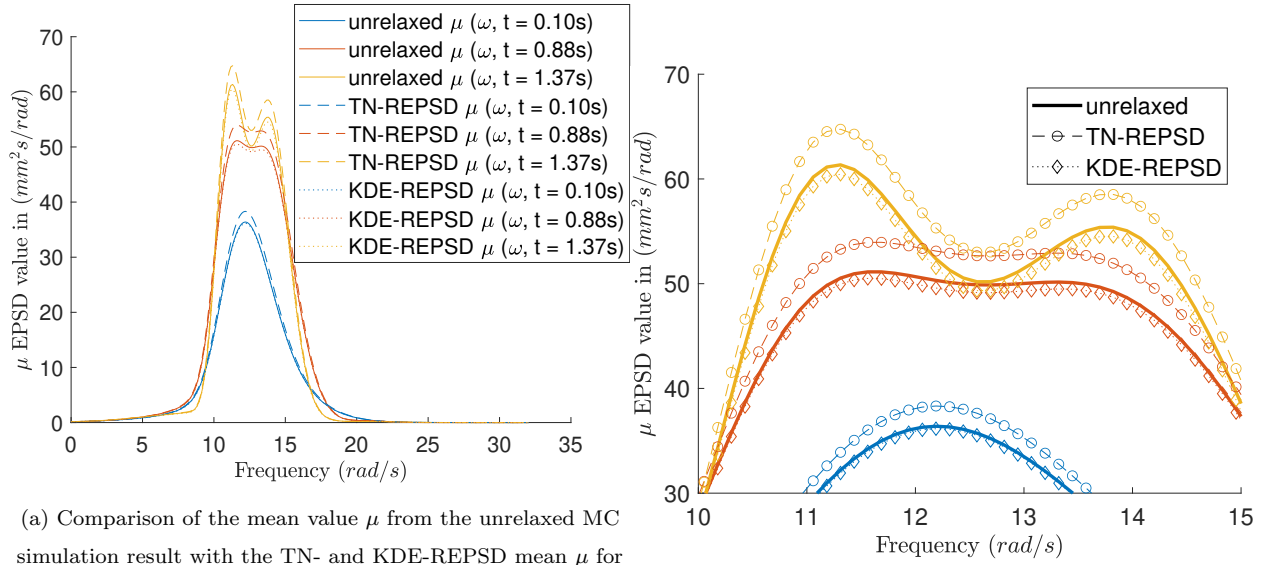


Figure 13: Full response statistics of each 5000 realised displacement histories of the FEM (both figures share the same legend)

further analyze the accuracy of the REPSD representation, for specific time steps the mean  $\mu$  in Fig. 14 and standard deviation  $\sigma$  in Fig. 15, for the unrelaxed MC EPSD and the TN- as well as the KDE-REPSD model are regarded. Unrelaxed refers to the fact, that no stochastic model for the EPSD function is considered. Additional stochastic signals are generated via MC with the respective REPSD representation, which would be impossible if only data or records of signals were available.

In Fig. 14(a) the mean value of 5000 MC realisations and their unrelaxed MTST estimation of the response EPSD is compared with the constructed TN- and KDE-REPSD functions. Please remember that the constructed REPSD representations are estimated out of 50 ensemble members. The mean value is compared for different time instants over the full frequency range. The different time instants are depicted in a different colour. From the results in Fig. 14(a) it can be seen that the TN-REPSD function seems to overestimate the benchmark EPSD function. This is specifically true for larger EPSD function values. The KDE-REPSD representation yields a better fit. For a closer comparison the frequency interval  $\omega \in [10, 15]$  rad/s is regarded in Fig. 14(b). These results also show that the KDE-REPSD does yield a better representation of the mean. Here in particular for the peak



(a) Comparison of the mean value  $\mu$  from the unrelaxed MC simulation result with the TN- and KDE-REPSD mean  $\mu$  for different time instances.

(b) Close up for the peak frequencies mean values.

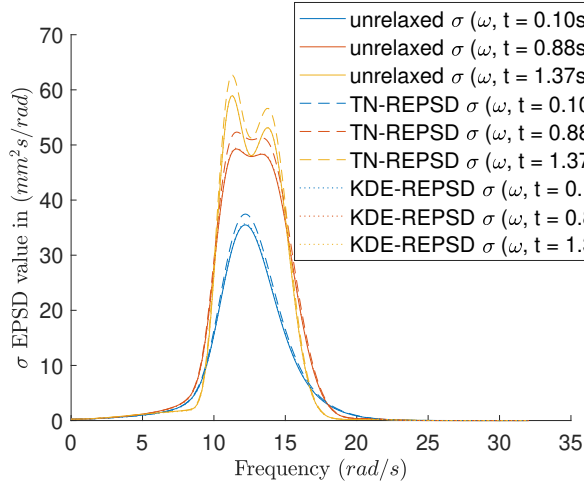
Figure 14: Comparison of mean EPSD values obtained from direct MC simulation (unrelaxed) and TN- and KDE-REPSD representation over the frequency for different time instances of interest.

EPSD values.

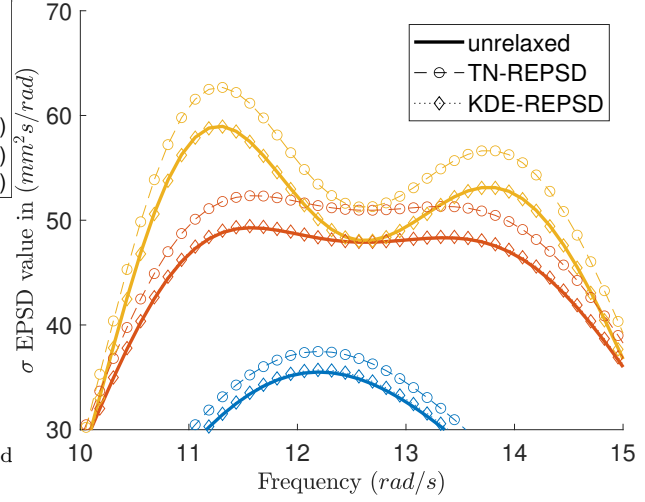
Analogous to the analysis of the EPSD functions mean value, the standard deviation  $\sigma$  is analyzed in Fig. 15. Different time instants are regarded and the corresponding EPSD function values over the frequency space are shown. The full frequency space is depicted in Fig. 15(a), where again the TN-REPSD representation is overestimating peak values of the response EPSD function standard deviation. For a closer comparison a smaller frequency interval can be seen in Fig. 15(b). These results indicate that for the standard deviation of the unrelaxed EPSD functions in comparison to the TN- and KDE-REPSD functions, the KDE-REPSD representation can reproduce the second order moment accurately from an ensemble with just 50 members.

#### 4.2. Modulated Davenport's power spectral density function for time dependant fluctuating wind speed simulation

In this example, the empirical Davenport's spectrum for near ground wind velocities as in [60] is regarded and modulated to simulate a time-dependant change of parameters. To achieve this, the basic Davenport's PSD function is modulated by a time dependant term



(a) Comparison of the standard deviation  $\sigma$  from the unrelaxed MC simulation result with the TN- and KDE-REPSD standard deviation  $\sigma$  for different time instances.



(b) Close up for the peak frequencies.

Figure 15: Comparison of the standard deviation  $\sigma$  for different EPSPDs obtained from direct MC simulation (unrelaxed) and TN- and KDE-REPSD representation over the frequency for different time steps of interest.

443  $A(t)$ . Similar as in [61] Davenport's power spectral density function can be written as

$$S_X^D(\omega, t) = A(t)v_*^2 \frac{\left(\frac{1200}{2\pi A(t)U_{10}}\omega\right)^2}{|\omega| \left(1 + \left(\frac{1200}{2\pi A(t)U_{10}}\omega\right)^2\right)^{4/3}}, \quad (17)$$

444 where  $v_* = 1.691$  m/s is the shear velocity of the wind and  $U_{10} = 31.88$  m/s is the 10 – min  
445 average wind speed in 10 m height. These are empirically estimated parameters.

446 A crude approach is suggested now. The hypothesis is that these two quantities could  
447 change in time. They could also have a correlation if a change in time is assumed. Without  
448 respecting the full physics of this complex coupled process at this point and only for the  
449 purpose of generating a challenging benchmark problem, we can assume that a time mod-  
450 ulation function of these quantities exists. This time modulation is denoted as  $A(t)$  and  
451 formulated by the following function

$$A(t) = \left| \sin\left(\frac{1}{2}t\right) \right| + \frac{1}{2}. \quad (18)$$

452 With this simple relation, an oscillatory change of the shear velocity and the the 10 – min

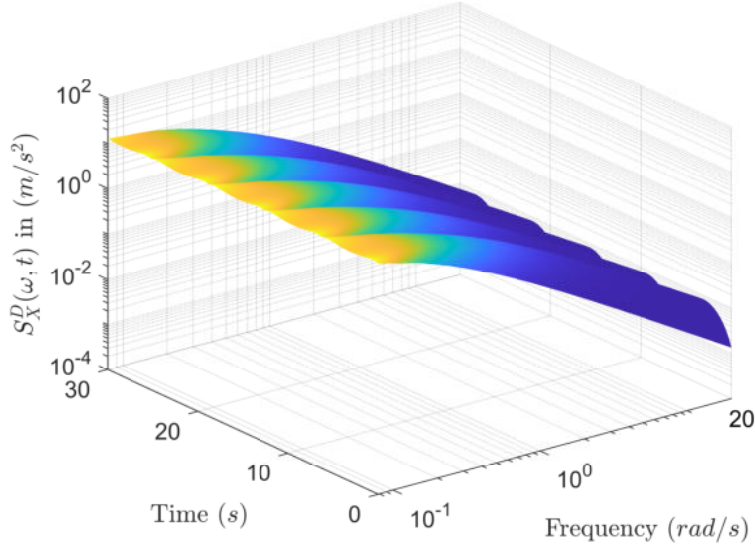


Figure 16: Modulated Davenport's separable EPSD function Eq. (3) for  $\omega \in [0, 20]$  rad/s and  $t \in [0, 30]$  s

average wind speed in 10m height is modelled. This modulation leads to a so-called separable EPSD function. This modulated Davenport's power spectral density function is denoted as  $S_X^D(\omega, t)$  and depicted in Fig. 16. Please note that this is an artificial benchmark problem, this modulation has not been tested or validated on any real processes connected to Davenport's PSD function. The resulting wind speed signals and their magnitude, are of course highly dependant on the modulation function. And in the spirit of the original Davenport's approach, should be validated through experiments empirically.

In this work, since the EPSD function is now fully analytically available, the SRM approach can be used to model stochastic processes that are non-homogeneous in time and frequency. Following parameters are used for the SRM and the domain specification: The total simulation time  $T^D = 30$  s, the upper cutoff-frequency  $\omega_u^D = 20$  rad/s, the number of discretised time steps  $N_t^D = 512$ , resulting in  $\Delta t^D = T^D/N_t^D$ . SRM yields again the artificial record set of 50 signals, depicted in Fig. 17. From these signals in Fig. 17, respectively the EPSD function ensemble members are estimated via the MTST approach. The same optimisation procedure as presented in Section 3.2 has been performed resulting in following MTST parameters:  $M^D = 2$ ,  $a^D = 0.4104$ ,  $b^D = 25.8864$ ,  $c^D = 0.3738$ . Three EPSD estimations

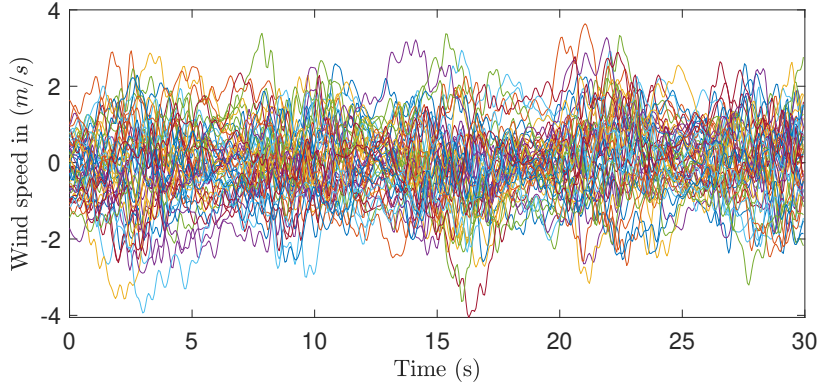
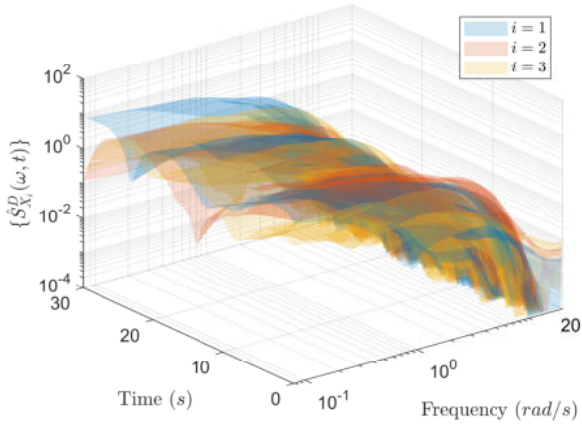


Figure 17: 50 artificial wind speed signals of the modulated Davenport's PSD (obtained by SRM)

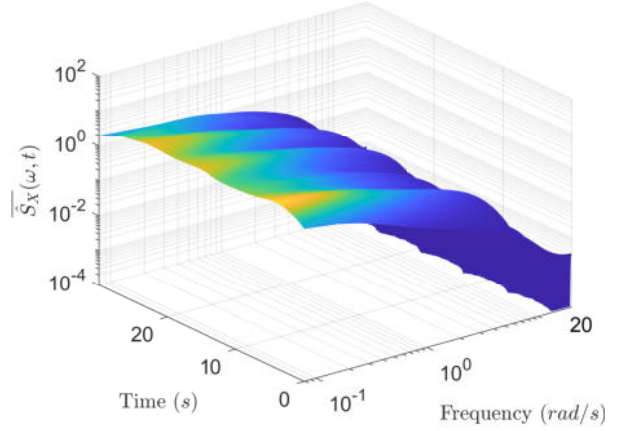
by MTST are depicted in Fig. 18(a) and the MTST estimation mean of the ensemble is represented in Fig. 18(b). The deterministic results as well as the statistical mean of the MTST estimations in Fig. 18 suggest that the variance and accuracy of the MTST estimation is limited. The source EPSD function in Eq. (17) represents a highly nonlinear relation, therefore the direct relation of the SRM and the MTST estimation of the EPSD functions can introduce further errors, which are not the research focus of this work. The modulated Davenport's PSD function can pose a challenging benchmark example for signal generation and EPSD function estimation procedures.

In this work, it is assumed that due to a limited amount of data, the estimation by MTST is the only information of the simulated signals in Fig. 17 that is present. However, applying other EPSD estimation procedures like Wavelet, STFT, Hilbert-Huang, or surrogate modelling techniques is beyond the scope of this work. This means, that the information present in Fig. 18 is considered to be the baseline. The hypothesis is, that for some natural processes like earthquakes, wind speed loading and sea wave loading, no source EPSD function is available or known. Only measurements and records are available.

The goal is now to establish an accurate relaxed representation of the EPSD function information within the ensemble, displayed in Fig. 18. This ensemble is hypothetically stemming purely from measurements and records. To achieve the relaxed representation of the EPSD function, first, the two presented REPSD representations TN-REPSD denoted by  $T\hat{S}_X^D(\omega, t)$  and KDE-REPSD, denoted by  $K\hat{S}_X^D(\omega, t)$  are constructed as discussed in the preliminar-



(a) Three deterministic EPSPD estimations by MTST, representative ensemble members



(b) Mean of the MTST EPSPD ensemble

Figure 18: The MTST estimation of the EPSPD for the by SRM and Eq. (17) generated wind speed signals in Fig. 17

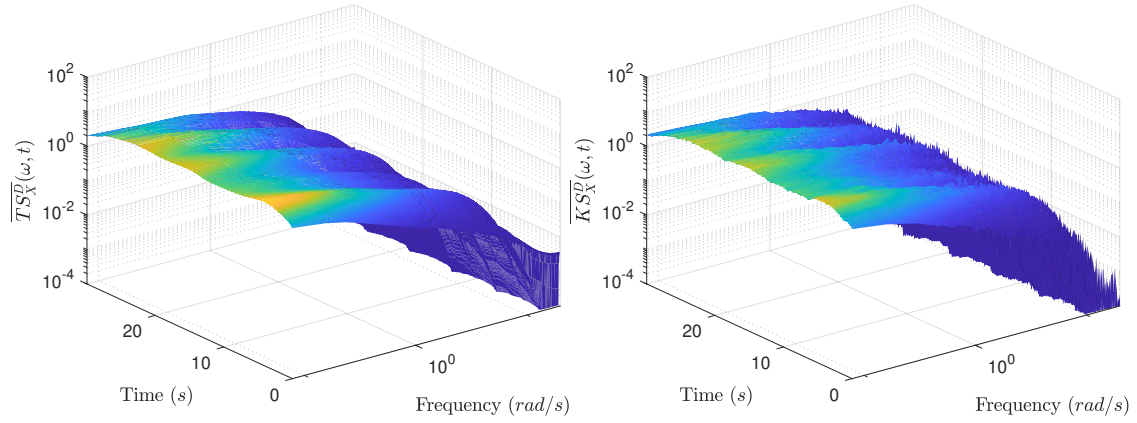
ies. Then the two representations are analysed for the accuracy of the MTST estimations. Therefore, a MC simulation is performed to generate 1000 new EPSPD samples from the respective REPSD functions. From these samples, the deviation towards the MTST estimation is analysed. These results are compiled in Fig. 19. Note that for the KDE approach in this example, a manual adjustment of the quartiles, introduced in Eq. (10), was necessary to reduce the modelling error:  $Q1 = 35.0646$  and  $Q2 = 45.0650$ , were chosen. In particular when looking at the mean of the 1000 generated MC samples for the TN-REPSD approach in Fig. 19(a) and the KDE-REPSD approach in Fig. 19(b) it is observable, that the TN-REPSD approach does model the mean of the ensemble better. The aforementioned adjustment of the quartiles was necessary because outliers of the ensemble were weighted too heavily into the establishment of the KDE. For the mean of the generated samples in Fig. 19(b), still some outliers can be identified. From this example it seems that the mean estimation of the TN-REPSD approach is more robust. This is also reflected in the error comparison. Here the mean of the samples for TN-REPSD are compared with the mean of the MTST estimated ensemble in Fig. 19(c), and the KDE-REPSD ensemble mean in Fig. 19(d). The KDE-REPSD representation does exhibit larger errors over the whole do-

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4 505 main. A contrary result can be observed for the standard deviation of the generated relaxed  
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6 506 samples in comparison to the standard deviation of the MTST EPSD function ensemble.  
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8 507 The TN-REPSD generated samples exhibit large areas of a larger error of the standard devi-  
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10 508 ation, see Fig. 19(e). The KDE-REPSD generated samples on the other hand, even though  
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12 509 having a similar maximum error range, exhibit errors in a smaller area Fig. 19(f). This  
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14 510 benchmark example highlights the importance of the choice of the underlying distribution  
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16 511 type for the relaxed representation of EPSD functions. Since for the TN-REPSD only the  
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18 512 mean value influences the parameter of the standard deviation (Eq. (11b)), no accurate rep-  
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20 513 resentation of varying data can be achieved. The KDE approach, with regard to an adjusted  
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22 514 IQR proofs to be more robust in terms of representing varying data.

23  
24 515 All in all it must be pointed out that any relaxed representation of EPSD functions is  
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26 516 highly dependant on the EPSD function estimation procedure. For the proposed approaches  
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28 517 in this example, the REPSD (of any distribution type) can only be as good as the EPSD  
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30 518 estimation procedure. But several techniques could be applied to improve the stochastic  
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32 519 model beneath the REPSD representations, such as Bayesian updating procedures, surrogate  
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34 520 representations or adaptive sampling approaches.

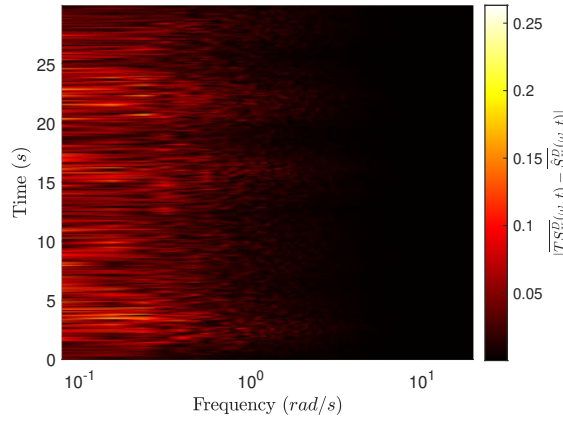
## 35 521 **5. Conclusions**

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38 522 In this work, a relaxed representation of EPSD functions for natural phenomena mod-  
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40 523 elled by stochastic processes has been proposed. The scope of this work was built around  
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42 524 the assumption, that for a battery of natural processes (e.g. seismic ground motion signals,  
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44 525 wind speed signals, wave load signals, or random vibrations in general) no prior informa-  
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46 526 tion about the source EPSD functions exists. However, data in the form of measurements  
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48 527 and records is available. It is possible to estimate EPSD functions deterministically by  
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50 528 time-frequency transformation methods, such as wavelet transformation, STFT, HHT, or  
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52 529 the recently developed MTST method. These transformations are only estimators and usu-  
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54 530 ally are underlying a large variance, especially for complex natural processes. The REPSD  
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56 531 function representation allows for the straightforward construction of a stochastic model  
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58 532 of the EPSD function, given data for the analysed processes is available. Not only the

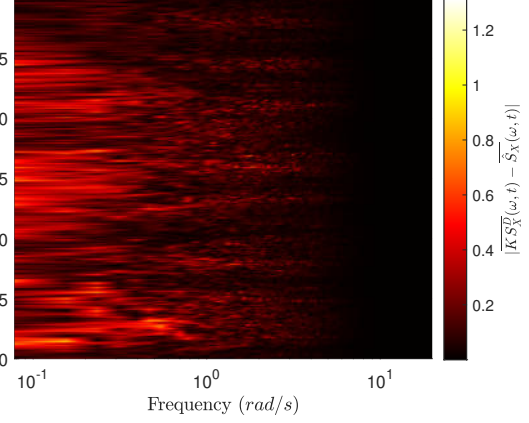


(a) TN-REPSD samples mean

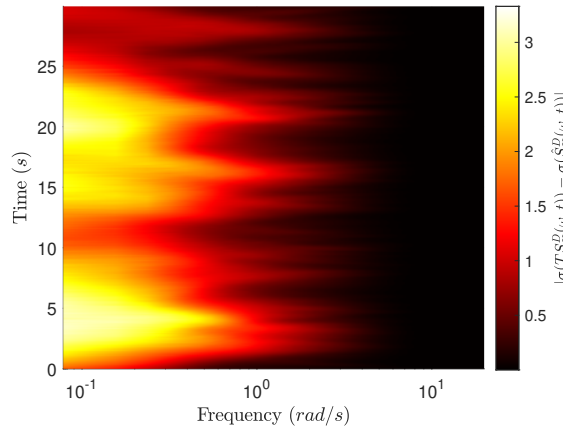
(b) KDE-REPSD samples mean



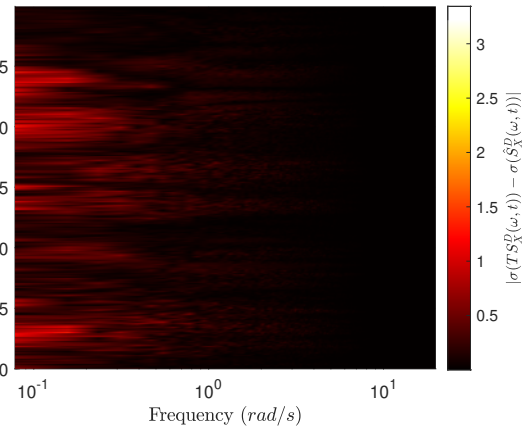
(c) Difference of TN-REPSD samples mean and MTST ensemble mean.



(d) Difference of KDE-REPSD samples mean and MTST ensemble mean.



(e) Difference of TN-REPSD samples standard deviation and MTST ensemble standard deviation  $\sigma$ .



(f) Difference of KDE-REPSD samples standard deviation and MTST ensemble standard deviation  $\sigma$ .

Figure 19: Error analysis of the relaxed representations of the MTST estimated EPD ensemble for the modulated Davenport's PSD.



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533 choice of time-frequency transformation method is modular, but also the underlying distri-  
534 bution type for the REPSD functions can be adaptively chosen to the specific problem at  
535 hand. For two simple approaches, a TN and KDE distribution type, this procedure has  
536 been presented in this work. In addition, it has been shown the the presented method is  
537 suitable for non-separable as well as separable EPSD functions. Convergence studies have  
538 been carried out to validate the results and analyse the accuracy. Additionally, the standard  
539 SRM method has been extended to a relaxed SRM approach, to be able to generate new  
540 stochastic signals, from the REPSD model. However, at several points in this work, it has  
541 been shown that the EPSD construction is not only dependant on the choice of the un-  
542 derlying distribution type, but is also heavily reliant on the accuracy of the time-frequency  
543 transformation method. An optimisation procedure for the MTST method has been pro-  
544 posed, which is applicable if further information on the EPSD function space is available.  
545 The REPSD formulation is applied to practical benchmark examples, where a FEM of an  
546 multiple degree of freedom system is analysed, here the input, as well as the response, are  
547 then modelled using the relaxed approach. Additionally, an academic benchmark example  
548 of a separable EPSD has been established, based on the empirical Davenport's PSD func-  
549 tion, which has been modulated to incorporate temporal features. The relaxed approach  
550 is used to construct a model that represents a relaxed stochastic EPSD for signals gener-  
551 ated by the modulated PSD. Overall, it can be stated that the REPSD representation's  
552 accuracy is as good as its underlying time-frequency transformation method. Nonetheless,  
553 the REPSD offers a stochastic representation of a stochastic processes model, already for  
554 a limited amount of data and records. The accuracy of these terms should further be val-  
555 idated by experiments and empirical investigations. Additionally, further information such  
556 as a correlation between the REPSD sample points could be considered. Also, a Bayesian  
557 updating procedure could be incorporated if the data set is of variable size, changing, or if  
558 additional parameters e.g. for the correlation are introduced. It is also possible to replace  
559 the probability distribution approach with surrogate modelling approaches, such as Gaus-  
560 sian process regression or Neural-Network-Representations. Thus, by the introduction of the  
561 REPSD concept, a generalised modular stochastic model for the representation of stochastic

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4 562 processes is established.

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7 563 **Declaration of competing interest**

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10 564 The authors declare that they have no known competing financial interests or personal  
11 565 relationships that could have appeared to influence the work reported in this paper.

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15 566 **CRedit author statement**

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18 567 **Marius Bittner:** Methodology, Formal analysis, Visualization, Validation, Software,  
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20 568 Writing - Original Draft. **Marco Behrendt:** Conceptualization, Methodology, Visualiza-  
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22 569 tion, Validation, Software, Writing - Review & Editing. **Michael Beer:** Project adminis-  
23  
24 570 tration, Supervision, Conceptualization, Funding acquisition, Writing - Review & Editing.

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## 10 11 761 **Appendix A. First and second order moments for matrices**

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14 762 Given the nature of the time-frequency transformation of stochastic processes, the avail-  
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16 763 able data of the EPSD functions in the set  $\{\hat{S}_{X_i}\}$  appear in matrix form. For the sake of  
17  
18 764 completeness here are the formal calculations of the first and second order moments. The  
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20 765 first order moment is given to be

$$21  
22  
23 \overline{S_X(\omega, t)} = \frac{1}{N_e} \sum_{i=1}^{N_e} S_{X_i}(\omega, t). \quad (\text{A.1})$$

24  
25  
26 766 The second order moment is given to be

$$27  
28  
29 \sigma(S_X(\omega, t)) = \sqrt{\frac{1}{N_e} \sum_{i=1}^{N_e} \left( S_{X_i}(\omega, t) - \overline{S_X(\omega, t)} \right)^2}. \quad (\text{A.2})$$

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